NOT RECOMMENDED FOR NEW DESIGNS intercil

na Sheet

EL5192, EL5192A

March 12, 2004

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FN7181.1
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600MHz Current Feedback Amplifier with Enable



The EL5192 and EL5192A are current feedback amplifiers with a very high bandwidth of 600MHz. This makes

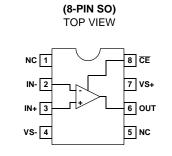
these amplifiers ideal for today's high speed video and monitor applications.

With a supply current of just 6mA and the ability to run from a single supply voltage from 5V to 10V, the amplifiers are also ideal for hand held, portable or battery-powered equipment.

The EL5192A also incorporates an enable and disable function to reduce the supply current to 100µA typical per amplifier. Allowing the \overline{CE} pin to float or applying a low logic level will enable the amplifier.

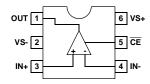
The EL5192 is offered in the 5-pin SOT-23 package and the EL5192A is available in the 6-pin SOT-23 as well as the industry-standard 8-pin SO packages. Both operate over the industrial temperature range of -40°C to +85°C.

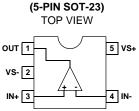
Pinouts



EL5192A

EL5192A (6-PIN SOT-23) TOP VIEW





EL5192

Features

- · 600MHz -3dB bandwidth
- · 6mA supply current
- Single and dual supply operation, from 5V to 10V supply span
- Fast enable/disable (EL5192A only)
- Available in SOT-23 packages
- Dual (EL5292) and triple (EL5392) available
- High speed, 1GHz product available (EL5191)
- Low power, 4mA, 300MHz product available (EL5193, EL5293, and EL5393)

Applications

- Video amplifiers
- Cable drivers
- RGB amplifiers
- Test equipment •
- Instrumentation
- · Current to voltage converters

Ordering Information

PART NUMBER	PACKAGE	TAPE & REEL	PKG. DWG. #
EL5192CW-T7	5-Pin SOT-23	7" (3K pcs)	MDP0038
EL5192CW-T7A	92CW-T7A 5-Pin SOT-23		MDP0038
EL5192ACW-T7	6-Pin SOT-23	7" (3K pcs)	MDP0038
EL5192ACW-T7A	6-Pin SOT-23	7"(250 pcs)	MDP0038
EL5192ACS	8-Pin SO	-	MDP0027
EL5192ACS-T7	8-Pin SO	7"	MDP0027
L5192ACS-T13 8-Pin SO		13"	MDP0027

Absolute Maximum Ratings (T_A = 25°C)

Supply Voltage between V _S + and V _S	.11V
Maximum Continuous Output Current	0mA
Operating Junction Temperature	25°C
Power Dissipation See Cu	urves

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

Electrical Specifications	V_{S} + = +5V, V_{S} - = -5V, R_{F} = 750 Ω for A_{V} = 1, R_{F} = 375 Ω for A_{V} = 2, R_{L} = 150 Ω , T_{A} = 25°C unless otherwise
	specified.

PARAMETER	DESCRIPTION	CONDITIONS	MIN	ТҮР	MAX	UNIT
AC PERFORM	ANCE	1	l	1	1	1
BW	-3dB Bandwidth	A _V = +1		600		MHz
		A _V = +2		300		MHz
BW1	0.1dB Bandwidth			25		MHz
SR	Slew Rate	$V_{O} = -2.5V$ to +2.5V, $A_{V} = +2$	2400	2800		V/µs
t _S	0.1% Settling Time	V_{OUT} = -2.5V to +2.5V, A_V = -1		9		ns
e _N	Input Voltage Noise			4.1		nV/√Hz
i _N -	IN- Input Current Noise			20		pA/√Hz
i _N +	IN+ Input Current Noise			50		pA/√Hz
dG	Differential Gain Error (Note 1)	A _V = +2		0.015		%
dP	Differential Phase Error (Note 1)	A _V = +2		0.04		0
DC PERFORM	ANCE				1	1
V _{OS}	Offset Voltage		-10	1	10	mV
T _C V _{OS}	Input Offset Voltage Temperature Coefficient	Measured from $\mathrm{T}_{\mathrm{MIN}}$ to $\mathrm{T}_{\mathrm{MAX}}$		5		µV/°C
R _{OL}	Transimpedance		200	400		kΩ
INPUT CHARA	CTERISTICS		<u> </u>			
CMIR	Common Mode Input Range		±3	±3.3		V
CMRR	Common Mode Rejection Ratio		42	50		dB
-ICMR	- Input Current Common Mode Rejection		-6		6	μA/V
+I _{IN}	+ Input Current		-60	3	60	μA
-I _{IN}	- Input Current		-35	2	35	μA
R _{IN}	Input Resistance			37		kΩ
C _{IN}	Input Capacitance			0.5		pF
OUTPUT CHAR	RACTERISTICS					
V _O	Output Voltage Swing	$R_L = 150\Omega$ to GND	±3.4	±3.7		V
		$R_L = 1k\Omega$ to GND	±3.8	±4.0		V
I _{OUT}	Output Current	$R_L = 10\Omega$ to GND	95	120		mA
SUPPLY						
I _{SON}	Supply Current - Enabled	No load, V _{IN} = 0V	5	6	7.5	mA
I _{SOFF}	Supply Current - Disabled	No load, V _{IN} = 0V		100	150	μA

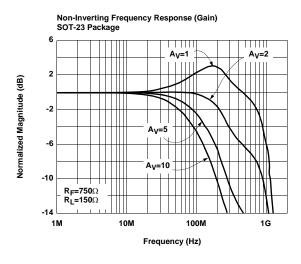
Electrical Specifications V_{S} + = +5V, V_{S} - = -5V, R_{F} = 750 Ω for A_{V} = 1, R_{F} = 375 Ω for A_{V} = 2, R_{L} = 150 Ω , T_{A} = 25°C unless otherwise specified. (Continued)

PARAMETER	DESCRIPTION	CONDITIONS	MIN	ТҮР	MAX	UNIT	
PSRR	Power Supply Rejection Ratio	DC, $V_{S} = \pm 4.75V$ to $\pm 5.25V$	55	75		dB	
-IPSR	- Input Current Power Supply Rejection	DC, $V_{S} = \pm 4.75V$ to $\pm 5.25V$	-2		2	µA/V	
ENABLE (EL51	I92A ONLY)	-	I				
t _{EN}	Enable Time			40		ns	
t _{DIS}	Disable Time			600		ns	
I _{IHCE}	CE Pin Input High Current	$\overline{CE} = V_S +$		0.8	6	μA	
I _{ILCE}	CE Pin Input Low Current	$\overline{CE} = V_{S}$ -		0	-0.1	μA	
VIHCE	CE Input High Voltage for Power- down		V _S + -1			V	
V _{ILCE}	CE Input Low Voltage for Power- down				V _S + -3	V	

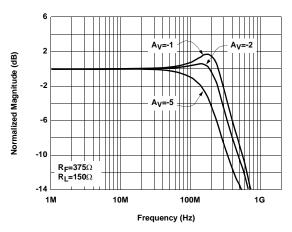
NOTE:

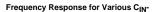
1. Standard NTSC test, AC signal amplitude = $286mV_{P-P}$, f = 3.58MHz

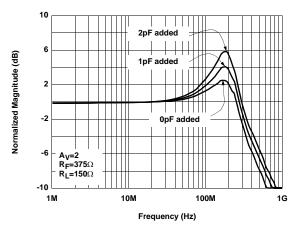
Typical Performance Curves



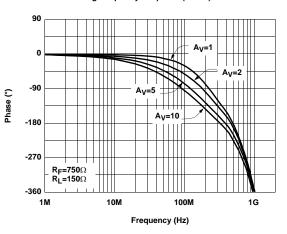
Inverting Frequency Response (Gain)



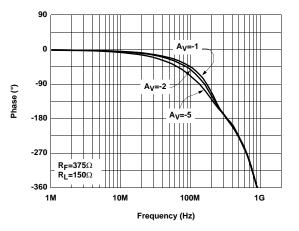


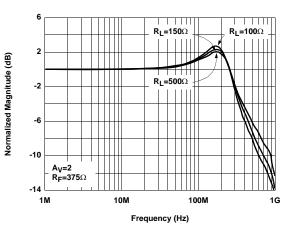


Non-Inverting Frequency Response (Phase)

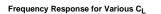


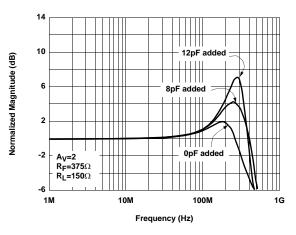
Inverting Frequency Response (Phase)

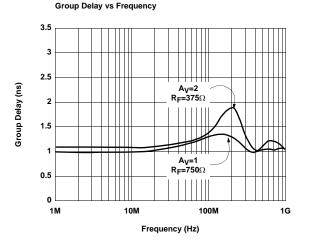




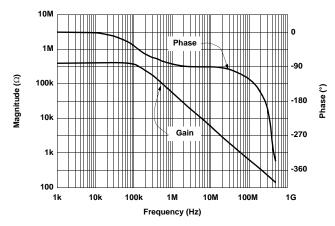
Frequency Response for Various RL



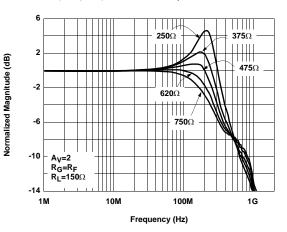


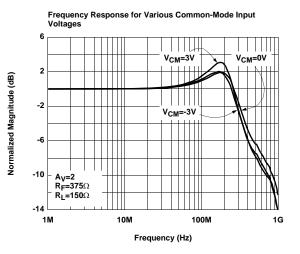




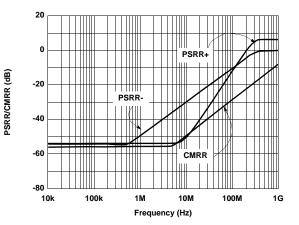


Frequency Response for Various R_F

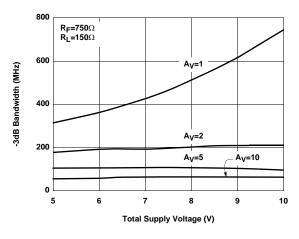




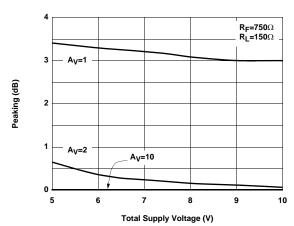


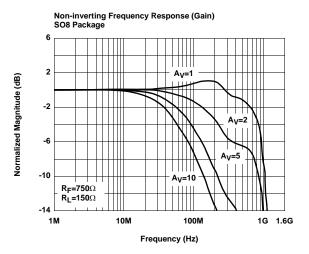


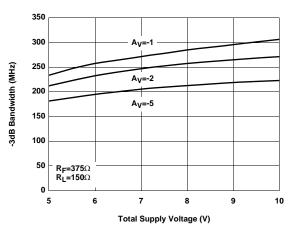
-3dB Bandwidth vs Supply Voltage for Non-Inverting Gains



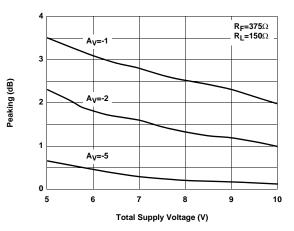
Peaking vs Supply Voltage for Non-Inverting Gains



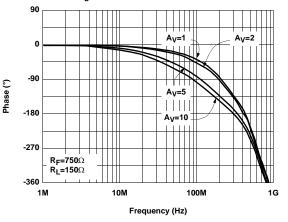




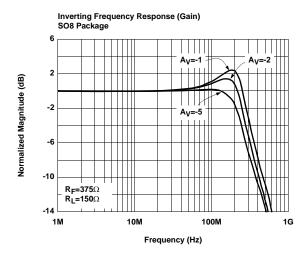
Peaking vs Supply Voltage for Inverting Gains



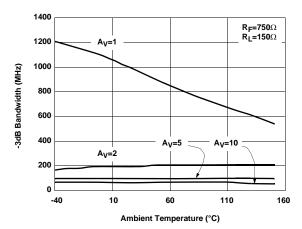
Non-inverting Frequency Response (Phase) SO8 Package



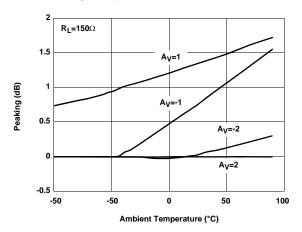
-3dB Bandwidth vs Supply Voltage for Inverting Gains

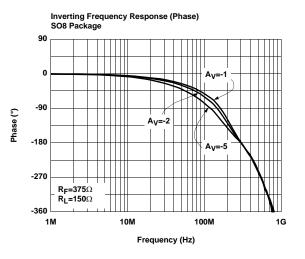


-3dB Bandwidth vs Temperature for Non-Inverting Gains

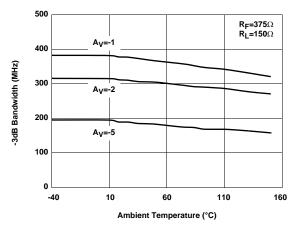


Peaking vs Temperature

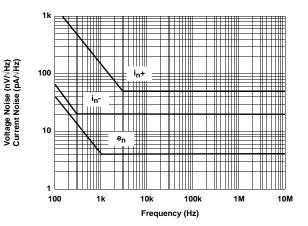


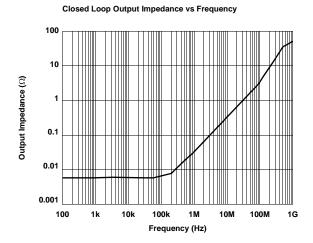


-3dB Bandwidth vs Temperature for Inverting Gains

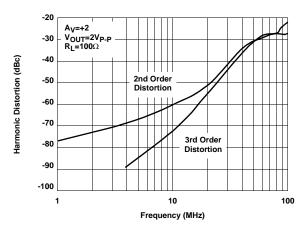


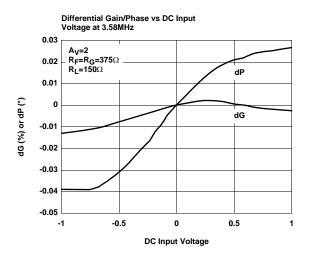


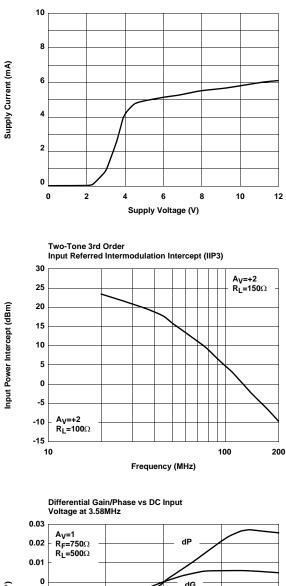




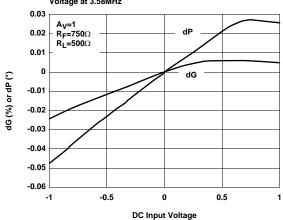
2nd and 3rd Harmonic Distortion vs Frequency

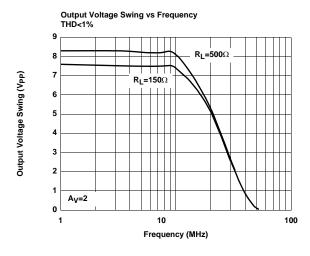






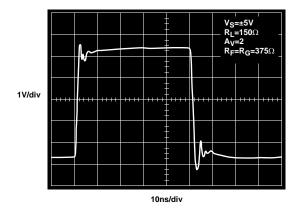
Supply Current vs Supply Voltage





Output Voltage Swing vs Frequency THD<0.1%

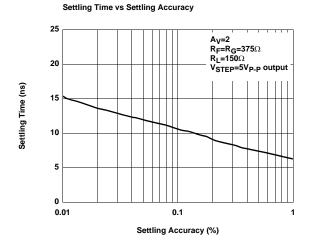
Large Signal Step Response



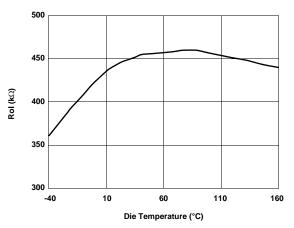
200mV/div

Small Signal Step Response

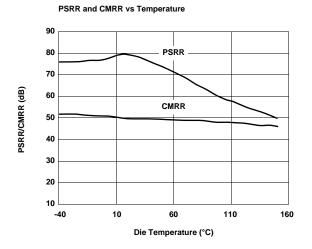
10ns/div

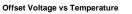


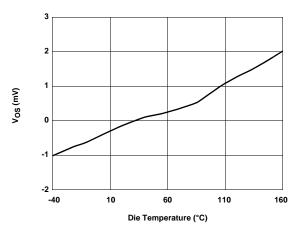


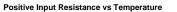


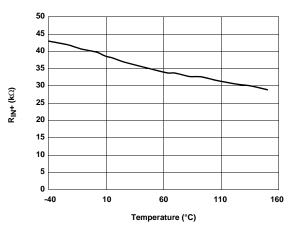
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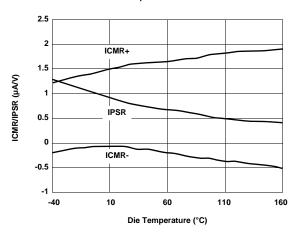


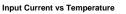


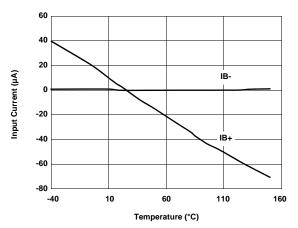


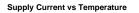


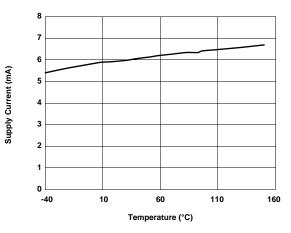
ICMR and IPSR vs Temperature

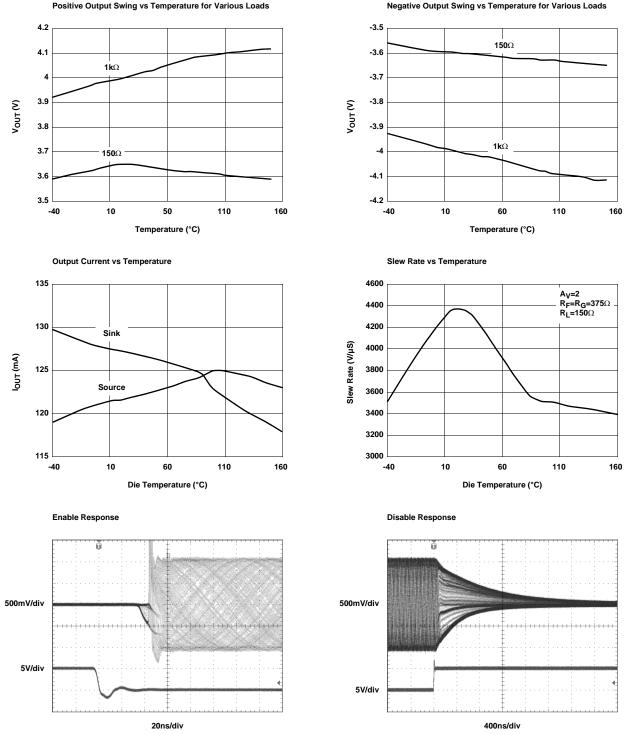




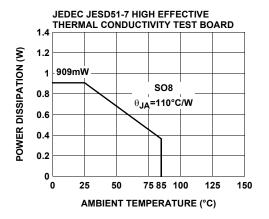


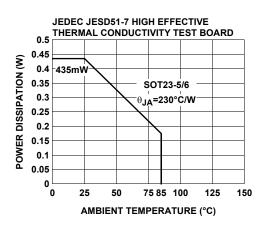


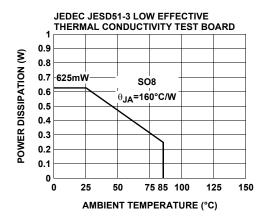


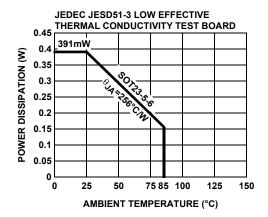


Negative Output Swing vs Temperature for Various Loads









Pin Descriptions

8-PIN SO	5-PIN SOT-23	6-PIN SOT-23	PIN NAME	FUNCTION	EQUIVALENT CIRCUIT
1, 5			NC	Not connected	
2	4	4	IN-	Inverting input	IN+ Circuit 1
3	3	3	IN+	Non-inverting input	(See circuit 1)
4	2	2	V _S -	Negative supply	
6	1	1	OUT	Output	V_{S^+}
7	5	6	V _S +	Positive supply	
8		5	CE	Chip enable	CE U VS+

Applications Information

Product Description

The EL5192 is a current-feedback operational amplifier that offers a wide -3dB bandwidth of 600MHz and a low supply current of 6mA per amplifier. The EL5192 works with supply voltages ranging from a single 5V to 10V and they are also capable of swinging to within 1V of either supply on the output. Because of their current-feedback topology, the EL5192 does not have the normal gain-bandwidth product associated with voltage-feedback operational amplifiers. Instead, its -3dB bandwidth to remain relatively constant as closed-loop gain is increased. This combination of high bandwidth and low power, together with aggressive pricing make the EL5192 the ideal choice for many low-power/highbandwidth applications such as portable, handheld, or battery-powered equipment.

For varying bandwidth needs, consider the EL5191 with 1GHz on a 9mA supply current or the EL5193 with 300MHz on a 4mA supply current. Versions include single, dual, and triple amp packages with 5-pin SOT-23, 16-pin QSOP, and 8-pin or 16-pin SO outlines.

Power Supply Bypassing and Printed Circuit Board Layout

As with any high frequency device, good printed circuit board layout is necessary for optimum performance. Low impedance ground plane construction is essential. Surface mount components are recommended, but if leaded components are used, lead lengths should be as short as possible. The power supply pins must be well bypassed to reduce the risk of oscillation. The combination of a 4.7μ F tantalum capacitor in parallel with a 0.01μ F capacitor has been shown to work well when placed at each supply pin.

For good AC performance, parasitic capacitance should be kept to a minimum, especially at the inverting input. (See the Capacitance at the Inverting Input section) Even when ground plane construction is used, it should be removed from the area near the inverting input to minimize any stray capacitance at that node. Carbon or Metal-Film resistors are acceptable with the Metal-Film resistors giving slightly less peaking and bandwidth because of additional series inductance. Use of sockets, particularly for the SO package, should be avoided if possible. Sockets add parasitic inductance and capacitance which will result in additional peaking and overshoot.

Disable/Power-Down

The EL5192A amplifier can be disabled placing its output in a high impedance state. When disabled, the amplifier supply current is reduced to < 150 μ A. The EL5192A is disabled when its \overline{CE} pin is pulled up to within 1V of the positive supply. Similarly, the amplifier is enabled by floating or pulling its \overline{CE} pin to at least 3V below the positive supply. For ±5V supply, this means that an EL5192A amplifier will be enabled when \overline{CE} is 2V or less, and disabled when \overline{CE} is above 4V. Although the logic levels are not standard TTL, this choice of logic voltages allows the EL5192A to be enabled by tying \overline{CE} to ground, even in 5V single supply applications. The \overline{CE} pin can be driven from CMOS outputs.

Capacitance at the Inverting Input

Any manufacturer's high-speed voltage- or current-feedback amplifier can be affected by stray capacitance at the inverting input. For inverting gains, this parasitic capacitance has little effect because the inverting input is a virtual ground, but for non-inverting gains, this capacitance (in conjunction with the feedback and gain resistors) creates a pole in the feedback path of the amplifier. This pole, if low enough in frequency, has the same destabilizing effect as a zero in the forward open-loop response. The use of largevalue feedback and gain resistors exacerbates the problem by further lowering the pole frequency (increasing the possibility of oscillation.)

The EL5192 has been optimized with a 375Ω feedback resistor. With the high bandwidth of these amplifiers, these resistor values might cause stability problems when combined with parasitic capacitance, thus ground plane is not recommended around the inverting input pin of the amplifier.

Feedback Resistor Values

The EL5192 has been designed and specified at a gain of +2 with R_F approximately 375 Ω . This value of feedback resistor gives 300MHz of -3dB bandwidth at A_V=2 with 2dB of peaking. With A_V=-2, an R_F of 375 Ω gives 275MHz of bandwidth with 1dB of peaking. Since the EL5192 is a current-feedback amplifier, it is also possible to change the value of R_F to get more bandwidth. As seen in the curve of Frequency Response for Various R_F and R_G, bandwidth and peaking can be easily modified by varying the value of the feedback resistor.

Because the EL5192 is a current-feedback amplifier, its gain-bandwidth product is not a constant for different closed-loop gains. This feature actually allows the EL5192 to maintain about the same -3dB bandwidth. As gain is increased, bandwidth decreases slightly while stability increases. Since the loop stability is improving with higher closed-loop gains, it becomes possible to reduce the value of R_F below the specified 375Ω and still retain stability, resulting in only a slight loss of bandwidth with increased closed-loop gain.

Supply Voltage Range and Single-Supply Operation

The EL5192 has been designed to operate with supply voltages having a span of greater than 5V and less than 10V. In practical terms, this means that the EL5192 will operate on dual supplies ranging from $\pm 2.5V$ to $\pm 5V$. With single-supply, the EL5192 will operate from 5V to 10V.

As supply voltages continue to decrease, it becomes necessary to provide input and output voltage ranges that can get as close as possible to the supply voltages. The EL5192 has an input range which extends to within 2V of either supply. So, for example, on \pm 5V supplies, the EL5192 has an input range which spans \pm 3V. The output range of the EL5192 is also quite large, extending to within 1V of the supply rail. On a \pm 5V supply, the output is therefore capable of swinging from -4V to +4V. Single-supply output range is larger because of the increased negative swing due to the external pull-down resistor to ground.

Video Performance

For good video performance, an amplifier is required to maintain the same output impedance and the same frequency response as DC levels are changed at the output. This is especially difficult when driving a standard video load of 150Ω , because of the change in output current with DC level. Previously, good differential gain could only be achieved by running high idle currents through the output transistors (to reduce variations in output impedance.) These currents were typically comparable to the entire 6mA supply current of each EL5192 amplifier. Special circuitry has been incorporated in the EL5192 to reduce the variation of output impedance with current output. This results in dG and dP specifications of 0.015% and 0.04°, while driving 150Ω at a gain of 2.

Video performance has also been measured with a 500Ω load at a gain of +1. Under these conditions, the EL5192 has dG and dP specifications of 0.03% and 0.05°, respectively.

Output Drive Capability

In spite of its low 6mA of supply current, the EL5192 is capable of providing a minimum of \pm 95mA of output current. With a minimum of \pm 95mA of output drive, the EL5192 is capable of driving 50 Ω loads to both rails, making it an excellent choice for driving isolation transformers in telecommunications applications.

Driving Cables and Capacitive Loads

When used as a cable driver, double termination is always recommended for reflection-free performance. For those applications, the back-termination series resistor will decouple the EL5192 from the cable and allow extensive capacitive drive. However, other applications may have high capacitive loads without a back-termination resistor. In these applications, a small series resistor (usually between 5 Ω and 50 Ω) can be placed in series with the output to eliminate most peaking. The gain resistor (R_G) can then be chosen to make up for any gain loss which may be created by this additional resistor at the output. In many cases it is also possible to simply increase the value of the feedback resistor (R_E) to reduce the peaking.

Current Limiting

The EL5192 has no internal current-limiting circuitry. If the output is shorted, it is possible to exceed the Absolute Maximum Rating for output current or power dissipation, potentially resulting in the destruction of the device.

Power Dissipation

With the high output drive capability of the EL5192, it is possible to exceed the 125°C Absolute Maximum junction temperature under certain very high load current conditions. Generally speaking when R_L falls below about 25 Ω , it is important to calculate the maximum junction temperature (T_{JMAX}) for the application to determine if power supply voltages, load conditions, or package type need to be modified for the EL5192 to remain in the safe operating area. These parameters are calculated as follows:

$$T_{JMAX} = T_{MAX} + (\theta_{JA} \times n \times PD_{MAX})$$

where:

T_{MAX} = Maximum ambient temperature

 θ_{JA} = Thermal resistance of the package

n = Number of amplifiers in the package

PD_{MAX} = Maximum power dissipation of each amplifier in the package

PD_{MAX} for each amplifier can be calculated as follows:

$$\mathsf{PD}_{\mathsf{MAX}} = (2 \times \mathsf{V}_{\mathsf{S}} \times \mathsf{I}_{\mathsf{SMAX}}) + \left[(\mathsf{V}_{\mathsf{S}} - \mathsf{V}_{\mathsf{OUTMAX}}) \times \frac{\mathsf{V}_{\mathsf{OUTMAX}}}{\mathsf{R}_{\mathsf{L}}} \right]$$

where:

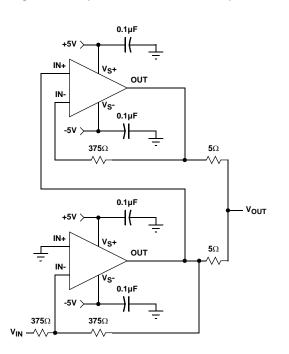
V_S = Supply voltage

I_{SMAX} = Maximum supply current of 1A

V_{OUTMAX} = Maximum output voltage (required)

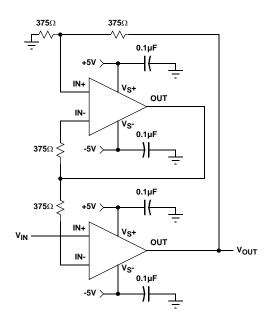
R_L = Load resistance

Typical Application Circuits



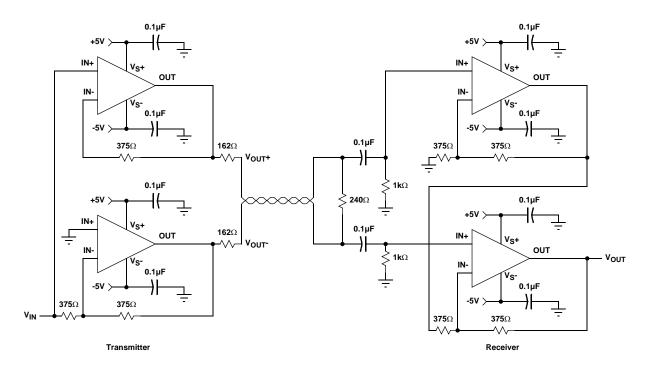
Inverting 200mA Output Current Distribution Amplifier

Fast-Settling Precision Amplifier



Typical Application Circuits (Continued)

Differential Line Driver/Receiver



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